

NCV7728

Hex Half-Bridge Driver

The NCV7728 is a Hex Half-Bridge Driver with protection features designed specifically for automotive and industrial motion control applications. The NCV7728 has independent controls and diagnostics. The device can be operated in forward, reverse, brake, and high impedance states. The drivers are controlled via a 16 bit SPI interface and are daisy chain compatible.

Features

- Low Quiescent Current Sleep Mode
- High-Side and Low-Side Drivers Connected in a Half-Bridge Configuration
- Integrated Freewheeling Protection (LS and HS)
- 0.55 A Peak Current
- $R_{DS(on)} = 1 \Omega$ (typ)
- 5 MHz SPI Control
- Compliance with 5 V and 3.3 V Systems
- Undervoltage and Overvoltage Lockout
- Discriminated Fault Reporting
- Overcurrent Protection
- Overtemperature Protection
- Under Load Detection (HS and LS)
- Daisy Chain Compatible with Multiple of 8 bit Devices
- 16-Bit Frame Detection
- NCV Prefix for Automotive and Other Applications Requiring Unique Site and Control Change Requirements; AEC-Q100 Qualified and PPAP Capable
- These are Pb-Free Devices

Typical Applications

- Automotive
- Industrial
- DC Motor Management for HVAC Application



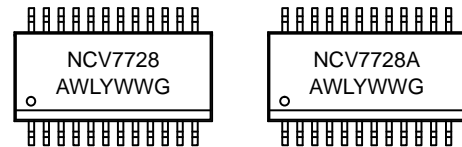
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**SSOP24 NB
CASE 565AL**

MARKING DIAGRAM



NCV7728 or = Specific Device Code
NCV7728A
A = Assembly Location
WL = Wafer Lot
Y = Year
WW = Work Week
G = Pb-Free Package

ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 26 of this data sheet.

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Shown below is a typical application for the NCV7728 configuration.

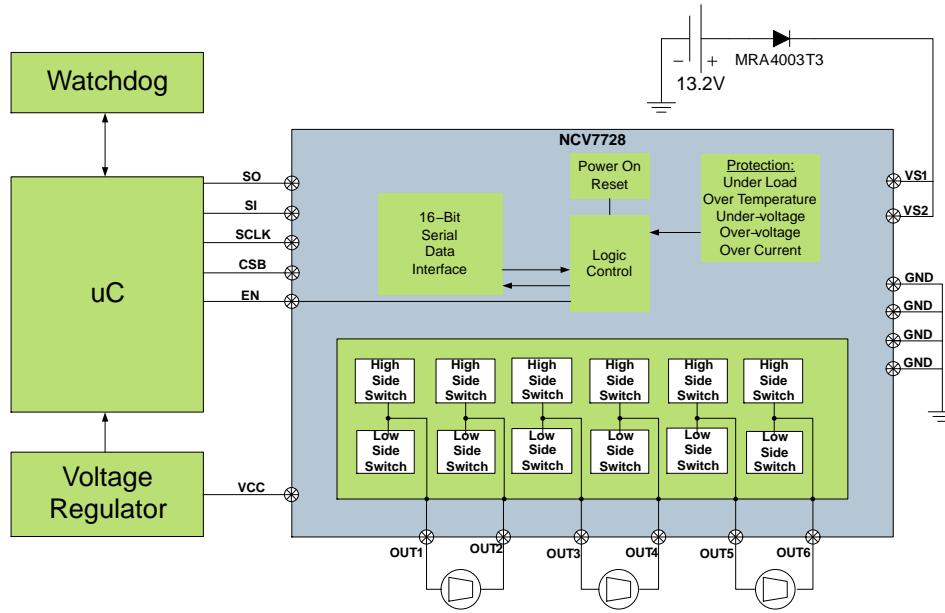


Figure 1. Typical Application

NCV7728

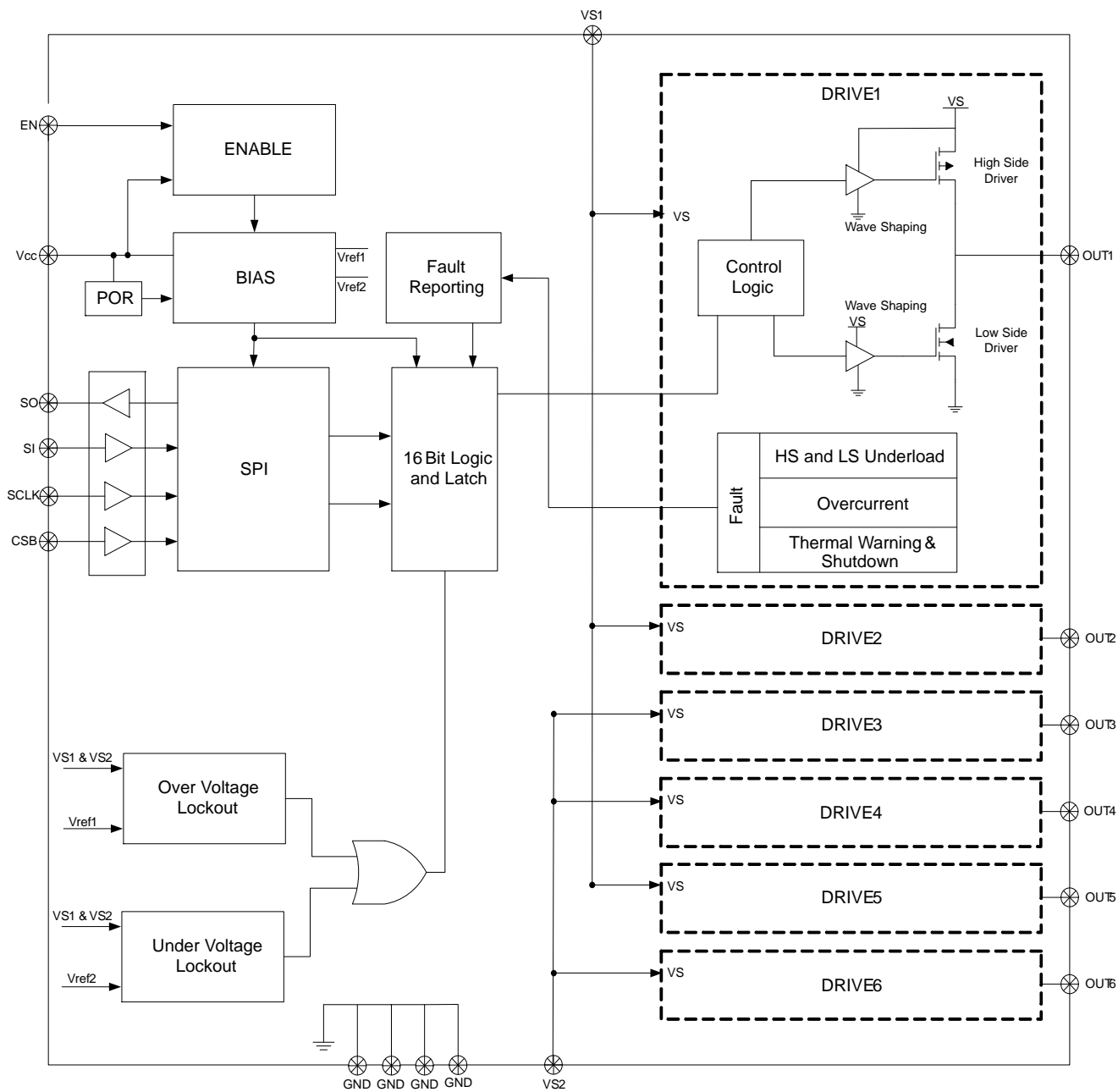


Figure 2. Block Diagram

NCV7728

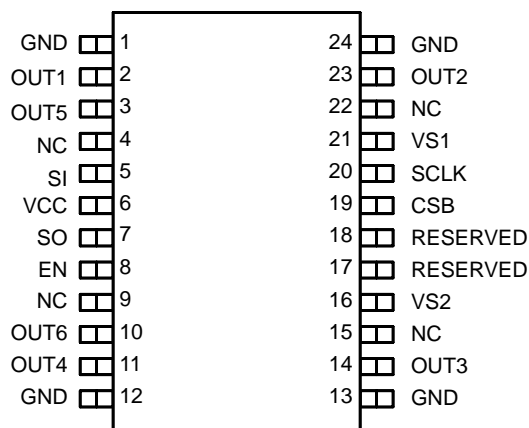


Figure 3. Pinout – SSOP24

PACKAGE DESCRIPTION The pin-out for the Hex Half-Bridge in SSOP24 package is shown in the table below.

Pin # SSOP24	Symbol	Description
1	GND	Ground. Shorted to pin 24 internally.
2	OUT1	Half Bridge Output 1
3	OUT5	Half Bridge Output 5
4	NC	No Connection. This pin should be isolated from any traces or via on the PCB board.
5	SI	Serial Input. 16 bit serial communications input. 3.3 V/5 V (TTL) Compatible. Internally pulled down.
6	VCC	Power supply input for Logic.
7	SO	Serial Output. 16 bit serial communications output. 3.3 V/5 V Complaint
8	EN	Enable. Input high wakes the IC up from a sleep mode. 3.3 V/5 V (TTL) Compatible. Internally pulled down.
9	NC	No Connection. This pin should be isolated from any traces or via on the PCB board.
10	OUT6	Half Bridge Output 6
11	OUT4	Half Bridge Output 4
12	GND	Ground. Shorted to pin 13 internally.
13	GND	Ground. Shorted to pin 12 internally.
14	OUT3	Half Bridge Output 3
15	NC	No Connection. This pin should be isolated from any traces or via on the PCB board.
16	VS2	Voltage Power Supply input for the Drivers 3, 4 and 6. This pin must be connected to VS1 externally.
17	Reserved	Factory use – connect to GND or leave unconnected – internally pulled down.
18	Reserved	Factory use – connect to GND or leave unconnected – internally pulled down.
19	CSB	Chip Select Bar. Active low serial port operation. 3.3V/5V (TTL) Compatible. Internally pulled up.
20	SCLK	Serial Clock. Clock input for use with SPI communication. 3.3 V/5 V (TTL) Compatible. Internally pulled down.
21	VS1	Voltage Power Supply input for the Drivers 1, 2 and 5, all the pre-drivers and the charge pump. This pin must be connected to VS2 externally.
22	NC	No Connection. This pin should be isolated from any traces or via on the PCB board.
23	OUT2	Half Bridge Output 2
24	GND	Ground. Shorted to pin 1 internally.

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MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Power Supply Voltage (VS1, VS2) (DC) (AC), t < 500ms, Ivsx > -2A	VsxdcMax VSXac	-0.3 to 40 -1.0	V
Output Pin OUTx (DC) (AC), t < 500ms, IOUTx > -1.1A (AC), t < 500ms, IOUTx < 1A	VoutxDc VoutxAc	-0.3 to 40 -1.0 1.0	V
Pin Voltage (Logic Input pins, SI, SCLK, CSB, SO, EN, VCC)	VioMax	-0.3 to 5.5	V
Output Current (OUT1, OUT2, OUT3, OUT4, OUT5, OUT6)	IoutxImax	-2.0 to 2.0	A
Electrostatic Discharge, Human Body Model, VSx, OUTx (AEC-Q100-002)	Vesd4k	4.0	kV
Electrostatic Discharge, Human Body Model, all other pins (AEC-Q100-002)	Vesd2k	2.0	kV
Electrostatic Discharge, Machine Model (AEC-Q100-003)	Vesd200	200	V
Short Circuit Reliability Characterization	AECQ10x	Grade A	-
Operating Junction Temperature	Tj	-40 to 150	°C
Storage Temperature Range	Tstr	-55 to 150	°C
Moisture Sensitivity Level (MAX 260°C Processing)	MSL2	2	-

Stresses exceeding those listed in the Maximum Ratings table may damage the device. If any of these limits are exceeded, device functionality should not be assumed, damage may occur and reliability may be affected.

THERMAL INFORMATION (Note 1)

Rating	Symbol	Value	Unit
Junction to Ambient	R _{θJA}	96.4	°C/W
Junction to Lead	R _{ψJL}	63	°C/W

1. Thermal Information is based on having 3 high side and 3 low side devices dissipating 80 mW each on a 2 layer board 0.062" thick FR4 board with 600 mm² copper spreader area. 2 oz copper is used for the copper spreader area and the ambient temperature is specified at 25°C.

RECOMMENDED OPERATING CONDITIONS

Rating	Symbol	Value		Unit
		Min	Max	
Digital Supply Input Voltage	VccOp	3.15	5.25	V
Battery Supply Input Voltage	VsxOp	5.5	28	V
DC Output Current	IxOp	-	0.55	A
Junction Temperature	TjOp	-40	125	°C

Functional operation above the stresses listed in the Recommended Operating Ranges is not implied. Extended exposure to stresses beyond the Recommended Operating Ranges limits may affect device reliability.

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ELECTRICAL CHARACTERISTICS ($-40^{\circ}\text{C} < T_J < 150^{\circ}\text{C}$, $5.5\text{ V} < V_{Sx} < 40\text{ V}$, $3.15\text{ V} < V_{CC} < 5.25\text{ V}$, $EN = V_{CC}$, unless otherwise specified)

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
GENERAL						
Supply Current ($V_{S1} + V_{S2}$) Sleep Mode	IqVsx85	$V_{S1} = V_{S2} = 13.2\text{ V}$, $V_{CC} = 0\text{ V}$ -40°C to 85°C No Load	–	1.0	2.5	μA
Supply Current ($V_{S1} + V_{S2}$) Active Mode	IvsOp	$EN = V_{CC}$, $5.5\text{ V} < V_{Sx} < 28\text{ V}$ No Load	–	2.5	5.0	mA
Supply Current (V_{CC}) Sleep Mode	IqV _{CC}	$CSB = V_{CC}$, $EN = SI = SCLK = 0\text{ V}$ (-40°C to 85°C)	–	1.0	2.5	μA
Active Mode	IV _{CC} Op	$EN = CSB = V_{CC}$, $SI = SCLK = 0\text{ V}$ No Load	–	1.5	3.0	mA
Total Sleep Mode Current $I(V_{S1}) + I(V_{S2}) + I(V_{CC})$	IqTot	Sleep Mode, -40°C to 85°C , No Load	–	2	5	μA
V_{CC} Power-On-Reset Threshold	V _{CC} por	V_{CC} increasing	–	2.55	2.9	V
V_{Sx} Undervoltage Detection Threshold	VsXuv	V_{Sx} decreasing	3.7	4.1	4.5	V
V_{Sx} Undervoltage Detection Hysteresis	VsXuHys		100	–	450	mV
V_{Sx} Overvoltage Detection Threshold	VsXov	V_{Sx} increasing	32	36	40	V
V_{Sx} Overvoltage Detection Hysteresis	VsXoHys		1	2.5	4	V

THERMAL RESPONSE

Thermal Warning	Twr	Not ATE tested	120	140	170	$^{\circ}\text{C}$
Thermal Warning Hysteresis	TwHy	Not ATE tested	–	20	–	$^{\circ}\text{C}$
Thermal Shutdown	Tsd	Not ATE tested	150	175	200	$^{\circ}\text{C}$
Thermal Shutdown Hysteresis	TsdHy	Not ATE tested	–	20	–	$^{\circ}\text{C}$

OUTPUTS

Output High $R_{DS(on)}$ (source)	RDSonHS	$I_{out} = -500\text{ mA}$, $V_{Sx} = 13.2\text{ V}$, $V_{CC} = 3.15\text{ V}$	–	1	2.25	Ω
Output Low $R_{DS(on)}$ (sink)	RDSonLS	$I_{out} = 500\text{ mA}$, $V_{Sx} = 13.2\text{ V}$, $V_{CC} = 3.15\text{ V}$	–	1	2.0	Ω
Output Path $R_{DS(HSx+LSx)}$	RDSonPath	$I_{out} = 500 \text{ mA}$	–	–	4.0	Ω
Source Leakage Current	IsrcLkg13.2 IsrcLkg28	$V_{CC} = 5\text{ V}$, $OUT(1-6) = 0\text{ V}$, -40°C to 85°C ; $V_{Sx} = 13.2\text{ V}$ $V_{Sx} = 28\text{ V}$	-1.0 -2.0	– –	– –	μA
Sink Leakage Current	I _{snk} Lkg13.2 I _{snk} Lkg28	$V_{CC} = 5\text{ V}$; $OUT(1-6) = V_{Sx} = 13.2\text{ V}$ $OUT(1-6) = V_{Sx} = 28\text{ V}$	– –	– –	1.0 2.0	μA
Overcurrent Shutdown Threshold (Source)	I _{sd} Src	$V_{CC} = 5\text{ V}$, $V_{Sx} = 13.2\text{ V}$	-2.0	-1.2	-0.8	A
Overcurrent Shutdown Threshold (Sink)	I _{sd} Snk	$V_{CC} = 5\text{ V}$, $V_{Sx} = 13.2\text{ V}$	0.8	1.2	2.0	A
Over Current Delay Timer	TdOc		10	25	50	μs
Under Load Detection Threshold (Low Side)	I _{uld} LS	$V_{CC} = 5\text{ V}$, $V_{Sx} = 13.2\text{ V}$	2.0	11	20	mA
Under Load Detection Threshold (High Side)	I _{uld} HS	$V_{CC} = 5\text{ V}$, $V_{Sx} = 13.2\text{ V}$	-20	-11	-2.0	mA
Under Load Detection Delay Time	TdUld	$V_{CC} = 5\text{ V}$, $V_{Sx} = 13.2\text{ V}$	200	350	600	μs

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ELECTRICAL CHARACTERISTICS ($-40^{\circ}\text{C} < T_J < 150^{\circ}\text{C}$, $5.5\text{ V} < V_{Sx} < 40\text{ V}$, $3.15\text{ V} < V_{CC} < 5.25\text{ V}$, $EN = V_{CC}$, unless otherwise specified)

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
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BODY DIODE

Power Transistor Body Diode Forward Voltage	VbdFwd	I _f = 500 mA	–	0.9	1.3	V
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LOGIC INPUTS (EN, SI, SCLK, CSB)

Input Threshold High Low	V _{thInH} V _{thInL}		2.0 –	– –	– 0.6	V
Input Hysteresis (SI, SCLK, CSB)	V _{thInHys}		50	150	300	mV
Enable Hysteresis	V _{thENHys}		150	400	800	mV
Input Pull-down Resistance (EN, SI, SCLK)	R _{pdx}	EN = SI = SCLK = V _{CC}	50	125	200	kΩ
Input Pull-up Resistance (CSB)	R _{puCSB}	CSB = 0 V	50	125	250	kΩ
Input Capacitance	C _{inx}	Not ATE tested	–	–	15	pF

LOGIC OUTPUT (SO)

Output High	V _{soH}	I _{SOURCE} = –1 mA	V _{CC} – 0.6	–	–	V
Output Low	V _{soL}	I _{SINK} = 1.6 mA	–	–	0.4	V
Tri-state Leakage	I _{triStLkg}	CSB = 5 V	–5	–	5	μA
Tri-state Output Capacitance	I _{triStCout}	CSB = V _{CC} , 0 V < V _{CC} < 5.25 V Not ATE tested	–	–	15	pF

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ELECTRICAL CHARACTERISTICS ($-40^{\circ}\text{C} < T_J < 150^{\circ}\text{C}$, $5.5\text{ V} < V_{Sx} < 40\text{ V}$, $3.15 < V_{CC} < 5.25\text{ V}$, $EN = V_{CC}$, unless otherwise specified)

Characteristic	Symbol	Conditions	Timing Chart	Min	Typ	Max	Unit
DRIVER OUTPUT TIMING SPECIFICATIONS							
High Side Turn On Time	ThsOn	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	7.5	13	μs
High Side Turn Off Time	ThsOff	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	3.0	6.0	μs
Low Side Turn On Time	TlsOn	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	6.5	13	μs
Low Side Turn Off Time	TlsOff	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	2.0	6.0	μs
High Side Rise Time	ThsTr	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	4.0	8.0	μs
High Side Fall Time	ThsTf	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	2.0	4.0	μs
Low Side Rise Time	TlsTr	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	1.0	3.0	μs
Low Side Fall Time	TlsTf	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		–	1.0	3.0	μs
High Side Off to Low Side On Non-Overlap Time	ThsOffLsOn	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		1.5	–	–	μs
Low Side Off to High Side On Non-Overlap Time	TlsOffHsOn	$V_{Sx} = 13.2\text{ V}$, $R_{load} = 39\ \Omega$		1.5	–	–	μs

SERIAL PERIPHERAL INTERFACE

SCLK Frequency	Fclk	$V_{CC} = 5\text{ V}$ $V_{CC} = 3.15\text{ V}$		– –	– –	5.0 2.0	MHz
SCLK Clock Period	TpClk	$V_{CC} = 5\text{ V}$ $V_{CC} = 3.15\text{ V}$		200 500	– –	– –	ns
SCLK High Time	TclkH		1	85	–	–	ns
SCLK Low Time	TclkL		2	85	–	–	ns
SCLK Setup Time	TclkSup		3 4	85 85	– –	– –	ns
SI Setup Time	TsiSup		11	50	–	–	ns
SI Hold Time	TsiH		12	50	–	–	ns
CSB Setup Time	TcsbSup		5 6	100 100	– –	– –	ns
CSB High Time (Note 2)	TcsbH		7	5.0	–	–	μs
SO enable after CSB falling edge	TenSo	$V_{CC} = 5\text{ V}$	8	–	–	200	ns
SO disable after CSB rising edge	TdisSo	$V_{CC} = 5\text{ V}$	9	–	–	200	ns
SO Rise Time	TsoR	$C_{load} = 40\text{ pF}$ Not ATE tested	–	–	10	25	ns
SO Fall Time	TsoF	$C_{load} = 40\text{ pF}$ Not ATE tested	–	–	10	25	ns
SO Valid Time	TsoV	$C_{load} = 40\text{ pF}$ SCLK \uparrow to SO 50%, Not ATE tested	10	–	20	50	ns
EN Low Valid Time (Note 3)	TenL	$V_{CC} = 5\text{ V}$ EN going low 50% to OUTx turning off 50%		10	–	–	μs
EN High to SPI Valid	TenHspiV			–	–	100	μs
SRR Delay Between Two Consecutive Frame (Note 4)	Tsrr			150	–	–	μs

2. This is the minimum time the user must wait between SPI commands.

3. This is the minimum time the user must wait before bringing EN up.

4. This is the minimum time the user must wait to send a SRR command between consecutive frames. If Tsrr time is not met the SRR request is ignored.

Product parametric performance is indicated in the Electrical Characteristics for the listed test conditions, unless otherwise noted. Product performance may not be indicated by the Electrical Characteristics if operated under different conditions.

ELECTRICAL CHARACTERISTIC TIMING DIAGRAMS

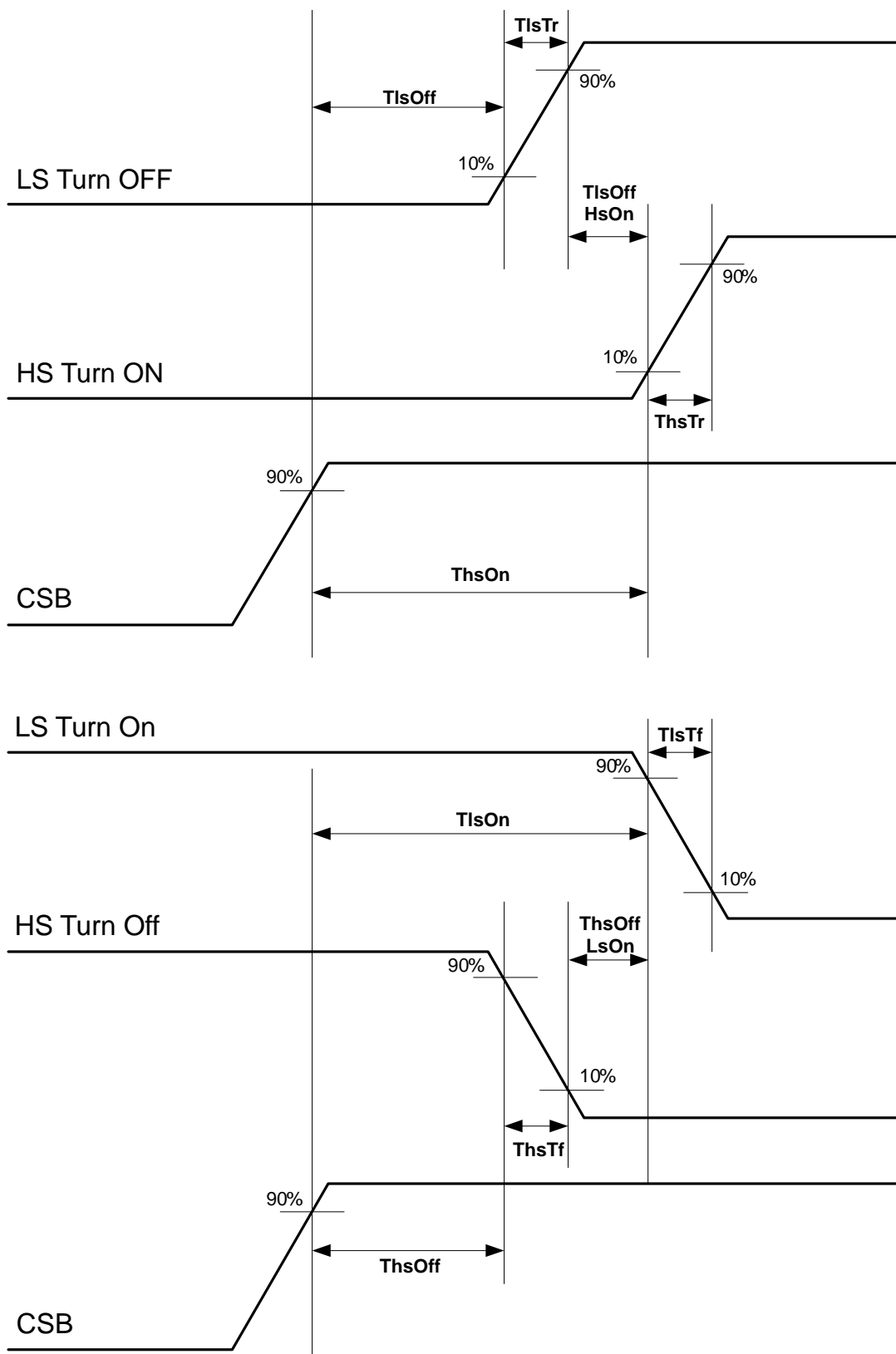


Figure 4. Detailed Driver Timing

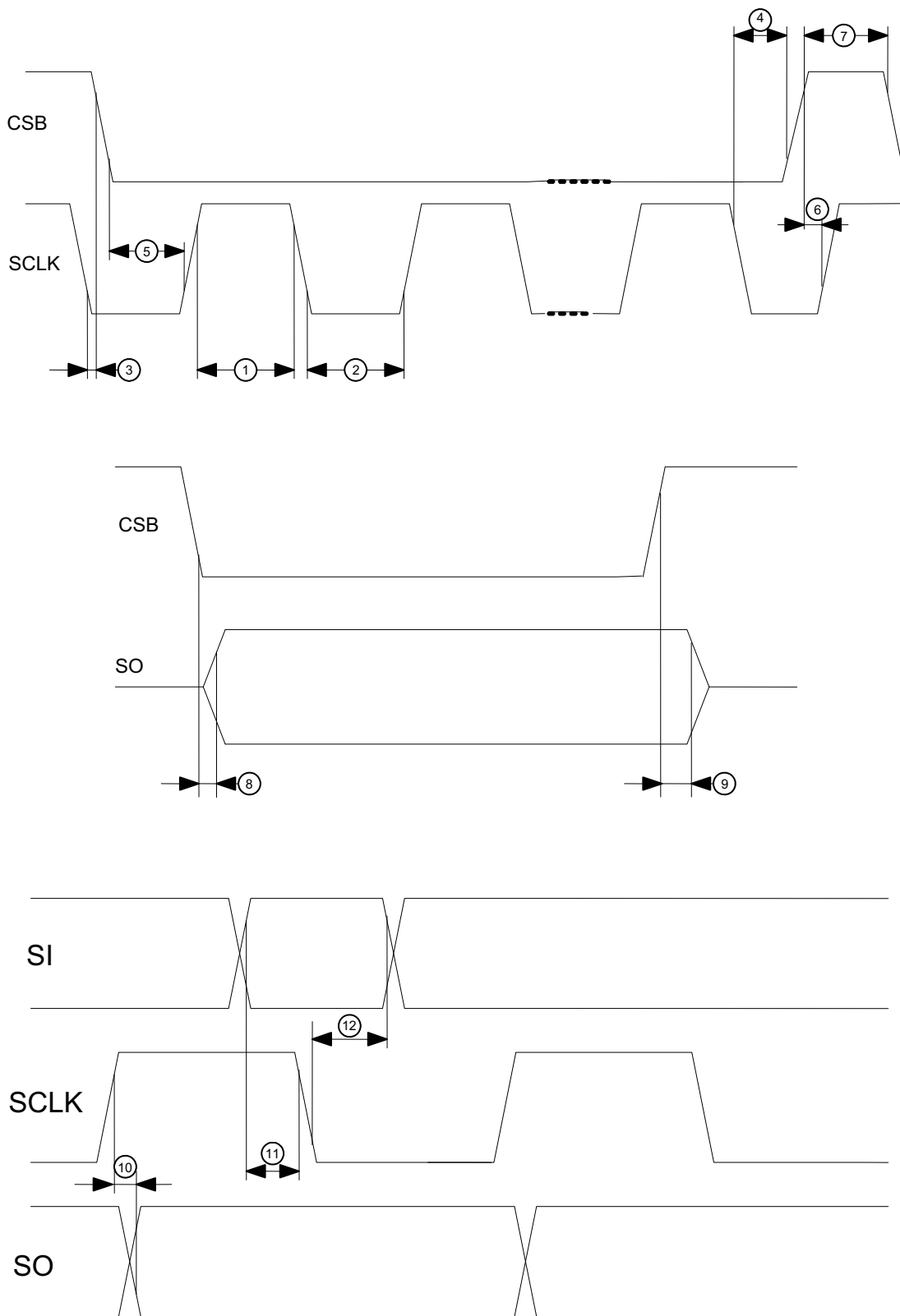


Figure 5. Detailed SPI Timing

TYPICAL PERFORMANCE GRAPHS

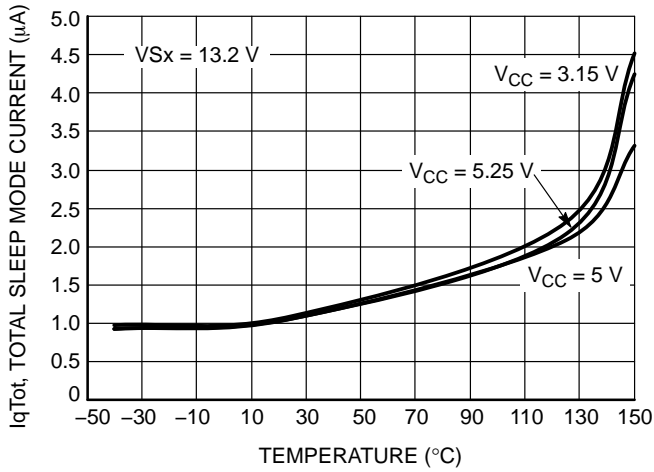


Figure 6. IqTot vs. Temperature

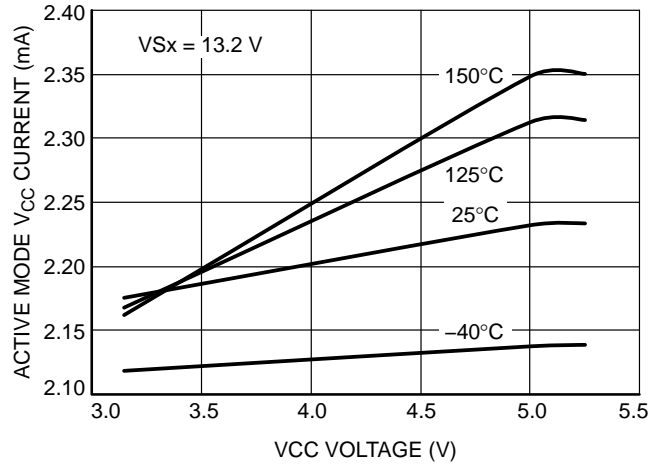


Figure 7. I(VCC) Active Mode vs. V(VCC)

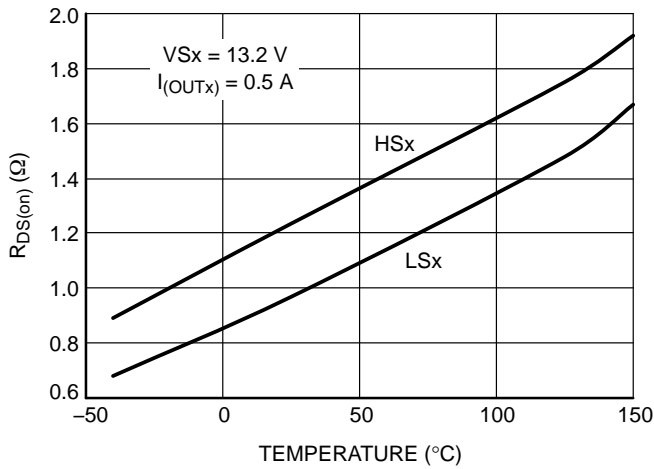


Figure 8. R_{DS(on)} vs. Temperature

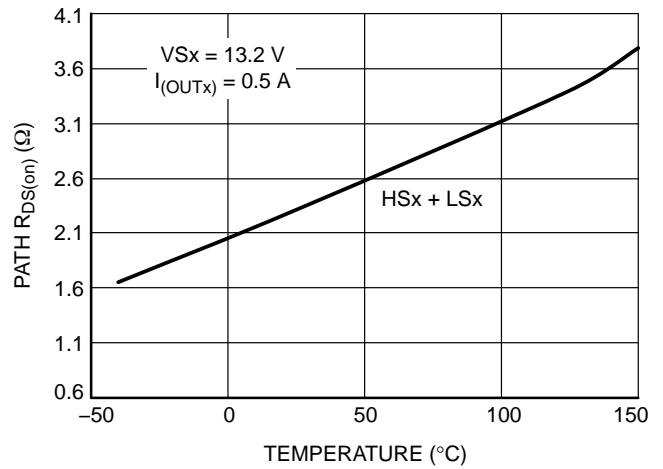


Figure 9. R_{DS(on)} vs. Temperature

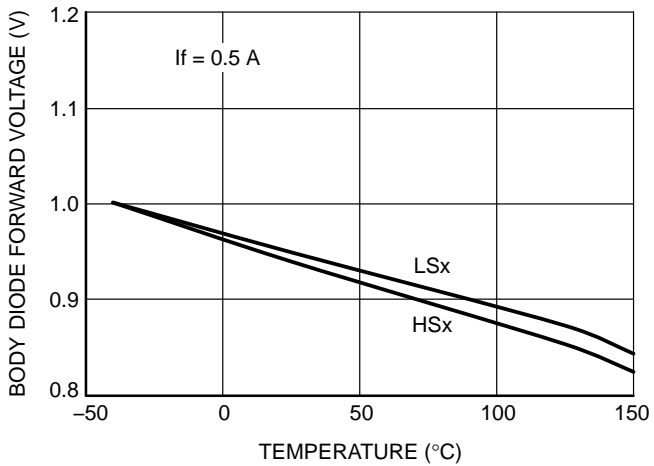


Figure 10. Body Diode vs. Temperature

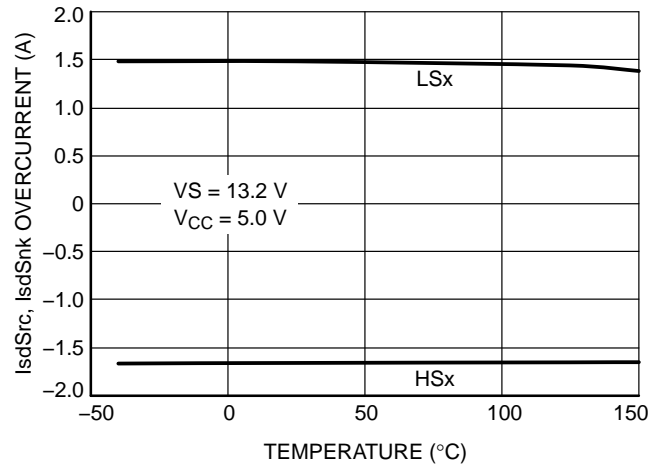


Figure 11. Overcurrent vs. Temperature

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TYPICAL PERFORMANCE GRAPHS

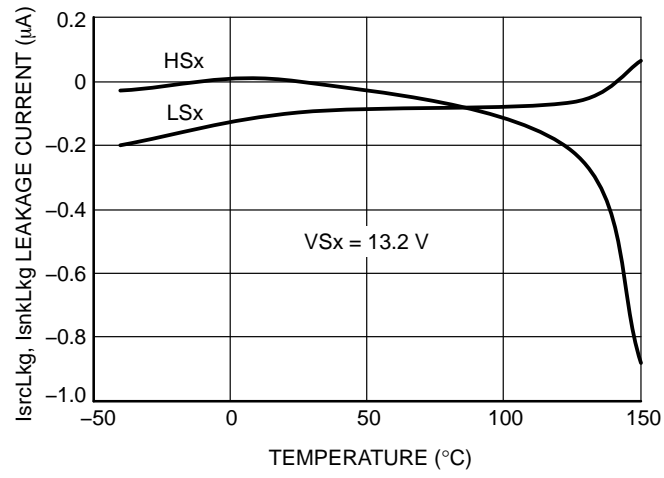


Figure 12. Source-Sink Leakage vs. Temperature

OPERATING DESCRIPTION

General Overview

The NCV7728 is comprised of twelve DMOS power drivers (six PMOS High Side Driver and six NMOS Low Side Driver) configured as six half bridges that enables three independent Full Bridge operations. Each output drive is characterized for a max 550 mA DC load and has a typical 2 A surge capability (at $V_{Sx} = 13.2$ V). Strict adherence to integrated circuit die temperature is necessary. Maximum die temperature is 150°C. This may limit the number of drivers enabled at one time. Output drive control and fault reporting is handled via the SPI (Serial Peripheral Interface) port.

An Enable function (EN) provides a low quiescent sleep current mode when the device is not being utilized. No data is stored when the device is in sleep mode. An internal pull down resistor is provided on the EN input to ensure the device is off if the input signal is lost. De-asserting the EN signal clears all the registers and resets the driver. When the EN signal is asserted the IC will proceed with the V_{CC} POR cycle and brings the drivers into normal operation.

SPI Communication

16-bit full duplex SPI communication has been implemented for the communication of this IC for device configurations, driver controls and reading the diagnostic data. In addition to the 16-bit diagnostic data, a pseudo bit (PRE_15) can also be retrieved from the SO register. The part is required to be enabled (EN active high) for SPI communication. The inputs for the SPI are TTL logic compatible and are specified by the V_{thInH} and V_{thInL} thresholds. The active low CSB input has a pull up resistor and the remaining SPI inputs have pull-down resistors to bias them to a known state when SPI is not active.

Reference the SPI communication frame format diagram in Figure 13 for the 16 bit SPI implementation. Tables 1 and 2 define the programming bits and diagnostic bits shown in Figure 13.

SPI COMMUNICATION FRAME FORMAT

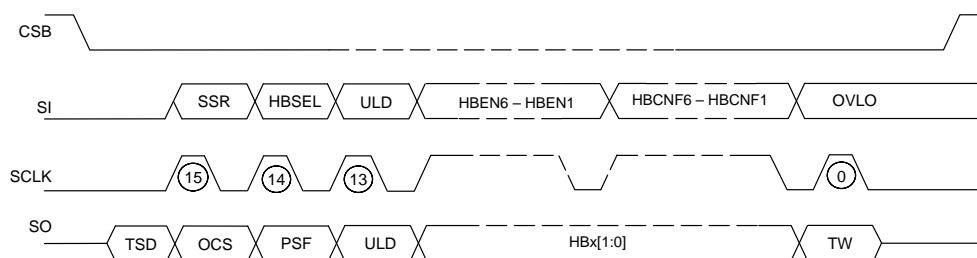


Figure 13. SPI Communication Frame Format

Communication is implemented as follows and is also illustrated in Figure 17:

1. SI and SCLK are set to low before the CSB cycle.
2. CSB goes low to allow serial data transfer.
3. SI data starting with the Most Significant bit (MSB) is shifted in first.
4. SI data is recognized on every falling edge of the clock.
5. Simultaneously, SO data from the previous frame starting with the MSB bit is shifted out on every rising edge of the clock.
6. The input data is compared to a 16 bit counter for the initial 16 bits shifted into SI for frame detection error scheme.
7. The sequential input bits are compared to a $n \times 8$ (n can take on the value of any integer) bit counter for daisy chain operations and are monitored by the frame detection error scheme.
8. CSB goes high and the most recent 16 bits clocked into SI are transferred to the data register given that there is no frame detection error. Otherwise the entire frame is ignored.
9. SO is tri-state when CSB is high.

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Table 1. SPI INPUT DATA FRAME

Input Data			
Bit Number	Bit Name	Bit Description	Bit Status
15	SRR	Status Reset Register When Asserted All Latched Faults are Cleared (TSD, OCS & ULD)	0 = No Reset
			1 = Reset
14	HBSEL (Note 5)	Half Bridge Selection	Reserved
13	ULDSC	Under Load Detection Shutdown Control Global Enable; Per Half Bridge Operation	0 = Disable
			1 = Enable
12	HBEN6	Half Bridge 6 Enable	0 = High Z
			1 = Enabled
11	HBEN5	Half Bridge 5 Enable	0 = High Z
			1 = Enabled
10	HBEN4	Half Bridge 4 Enable	0 = High Z
			1 = Enabled
9	HBEN3	Half Bridge 3 Enable	0 = High Z
			1 = Enabled
8	HBEN2	Half Bridge 2 Enable	0 = High Z
			1 = Enabled
7	HBEN1	Half Bridge 1 Enable	0 = High Z
			1 = Enabled
6	HBCNF6	Half Bridge 6 Configuration Control	0 = LS6 ON & HS6 OFF
			1 = LS6 OFF & HS6 ON
5	HBCNF5	Half Bridge 5 Configuration Control	0 = LS5 ON & HS5 OFF
			1 = LS5 OFF & HS5 ON
4	HBCNF4	Half Bridge 4 Configuration Control	0 = LS4 ON & HS4 OFF
			1 = LS4 OFF & HS4 ON
3	HBCNF3	Half Bridge 3 Configuration Control	0 = LS3 ON & HS3 OFF
			1 = LS3 OFF & HS3 ON
2	HBCNF2	Half Bridge 2 Configuration Control	0 = LS2 ON & HS2 OFF
			1 = LS2 OFF & HS2 ON
1	HBCNF1	Half Bridge 1 Configuration Control	0 = LS1 ON & HS1 OFF
			1 = LS1 OFF & HS1 ON
0	OVLO	Over Voltage Lock Out Global Effect	0 = Disable
			1 = Enable

5. HBSEL enables bridge selection for the NCV7719 and NCV7720 devices. In the NCV7728 it is recommended to set the HBSEL to zero.

Table 2. SPI OUTPUT DATA FRAME

Output Data				
Bit Number	Bit Name	Bit Description	Bit Status	
PRE_15	TSD	Latched Thermal Shutdown	0 = No Fault	
			1 = Fault	
15	OCS	Over Current Shutdown Global Notification	0 = No Fault	
			1 = Fault	
14	PSF	Power Supply Failure on VS1 and/or VS2 Under Voltage and Over Voltage Monitoring	0 = No Fault	
			1 = Fault	
13	ULD	Under Load Detection Global Notification	0 = No Fault	
			1 = Fault	
12	HB6 [1:0]	Half Bridge 6 Status	0x00b – Output Disable 0x01b – OCS 0x10b – ULD 0x11b – Output Enabled	
11				
10	HB5 [1:0]	Half Bridge 5 Status		
9				
8	HB4 [1:0]	Half Bridge 4 Status		
7				
6	HB3 [1:0]	Half Bridge 3 Status		
5				
4	HB2 [1:0]	Half Bridge 2 Status		
3				
2	HB1 [1:0]	Half Bridge 1 Status		
1				
0	TW	Thermal Warning Global Notification		0 = No Fault
				1 = Fault

HBx[1:0] bits are priority encoded to provide the status information of each of the half-bridge outputs. Figure 14 shows the priority encoding state diagram for the HBx[1:0] bits.

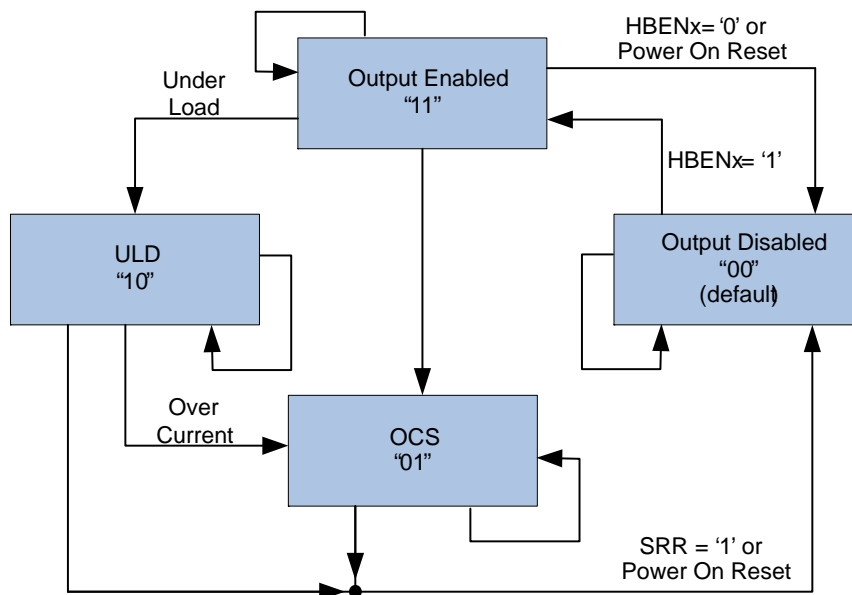


Figure 14. SO HBx[1:0] Priority Encoding State Diagram

As seen from Figure 14 an over current event following an under load condition on the same half-bridge reports HBx = '10' then HBx = '01'. However, an over current preceding an under load fault only reports HBx = '01' since there are no direct path from the OCS state to the ULD state. Thus an over current shutdown fault must be cleared before an under load fault is reported on the same half-bridge.

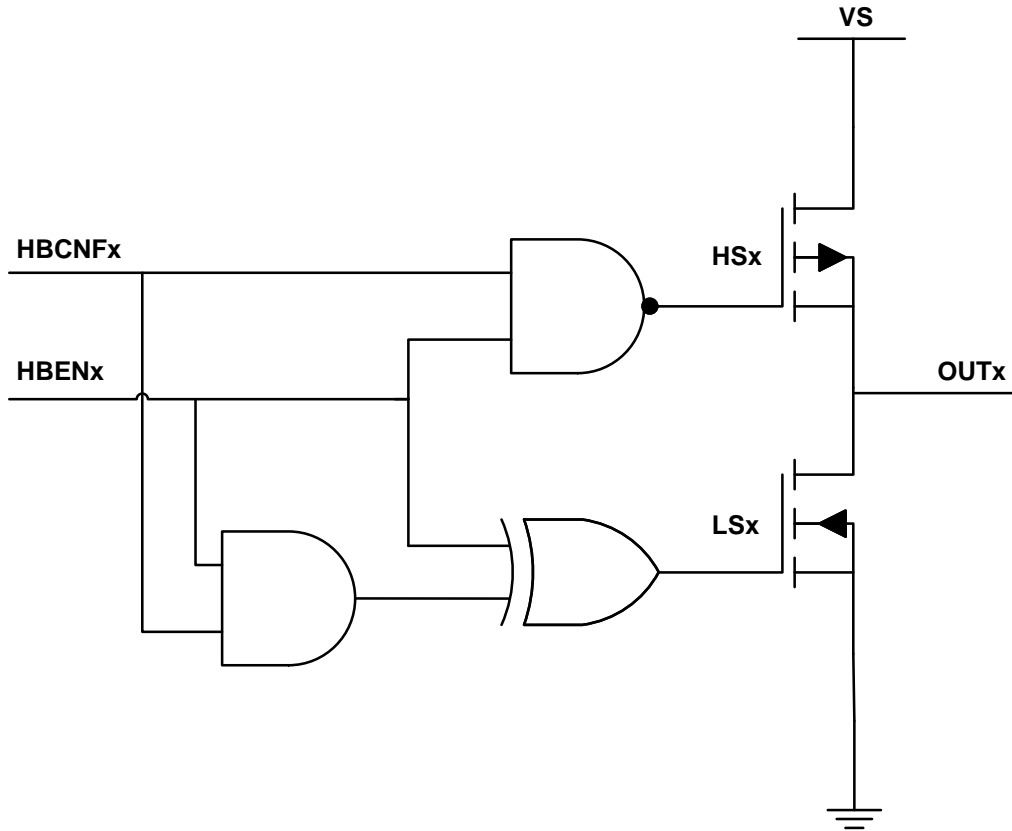
HBx[1:0] bits are priority encoded to provide the status information of each of the half-bridge outputs. Thus an over current shutdown fault must be cleared before an under load fault is reported on the same half-bridge. The latched thermal shutdown (TSD) information is available on SO after CSB goes low until the first rising SCLK edge. The following procedures must be met for a true TSD reading:

1. SCLK and SI are low before the CSB cycle. Violating these conditions will results in an undetermined SPI behavior or/and an incorrect TSD reading.
2. CSB transitioning from high to low.
3. CSB setup time (TcsbSup) is satisfied and the data is captured before the first SCLK rising edge.

Driver Control

The NCV7728 has the flexibility of controlling each driver through the 16 bit SPI frame (Bits 12–1) and the logic

combination required for bridge control is defined in Figure 15.



HBENx	HBCNFx	OUTx
0	'X'	OUTx in High Impedance State
1	0	HSx Off and LSx On
1	1	HSx On and LSx Off

'X' = Don't Care

Figure 15. Bridge Control Logic

The digital design insures that the high side and low side of the same half bridge will not be active at the same time. Thus the device self protects from a current shoot through condition. Delays (T_{hsOffLsOn} and T_{lsOffHsOn}) between the high side and low side switching are implemented for same reasons.

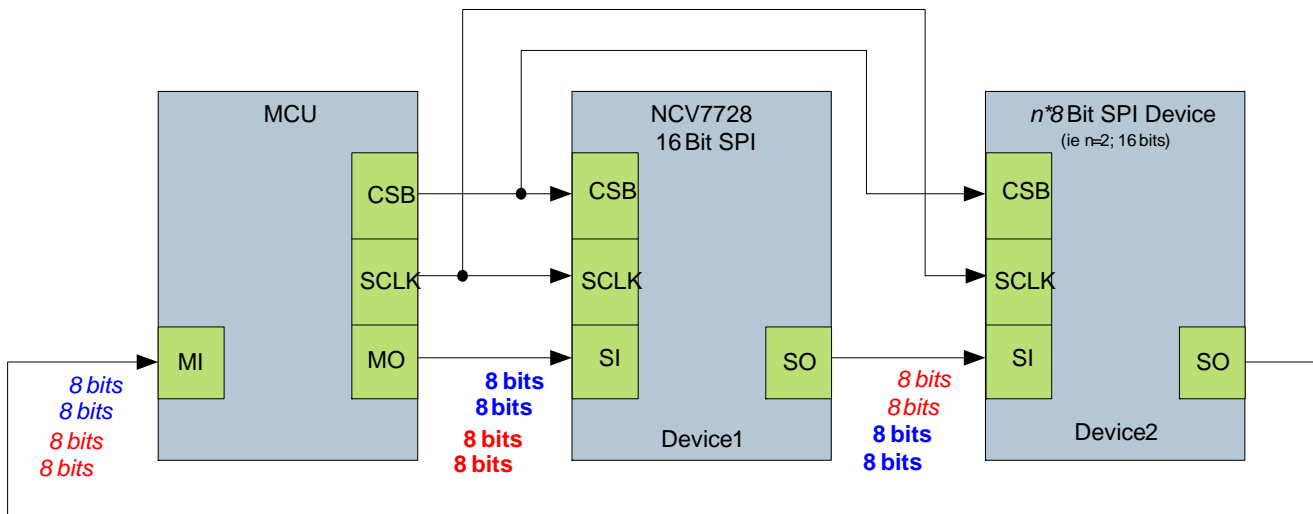
Frame Detection

To maintain the data integrity, the NCV7728 has 16 bit frame detection. A valid frame for a single CSB cycle requires 16 bits to be clocked into SI for the initial 16 bits and $n \times 8$ bits thereafter. In an instance of an invalid SPI frame the entire frame is ignored, but the previous states of the corresponding outputs are maintained.

Daisy Chain Operation

Daisy chain communications between multiple of 8-bit SPI compatible IC's is possible by connection of the serial output pin (SO) to the input of the sequential IC (SI). The clock phase and clock polarity respect to the data must be the same for all the devices on the chain. Figure 16 illustrates the hardware configuration of NCV7728 daisy chained with a $n \times 8$ bit (ie $n = 2$; 16 bit) SPI device. The progression of data from the MCU through the sequential devices is also shown. Strict adherence to the frame format illustrated in Figure 17 is required for the proper serial daisy chain operations.

NCV7728



Command Bits for the Device 2

Previous Diagnostic Bits from Device2

Command Bits for Device 1

Previous Diagnostic Bits From Device1

Figure 16. Serial Daisy Chain

If Device 2 is a 16 bit IC, then a total of 32 bits must be generated from the MCU for a complete transport of data in the system. Monitoring of all the devices in the serial chain must be employed on a system level architecture. Thus, pre-cautious measure should be taken to avoid situations where not enough frames were sent to the devices, but the frames transmitted did not violate the internal frame

detection counters. For these scenarios, invalid data is accepted by NCV7728 and possibly by other devices on the chain depending on their frame detection design. The data shifted in will be transferred to the data registers of the devices on the beginning of the chain and the devices at the end of the chain will get the previous diagnostic data of the preceding devices.

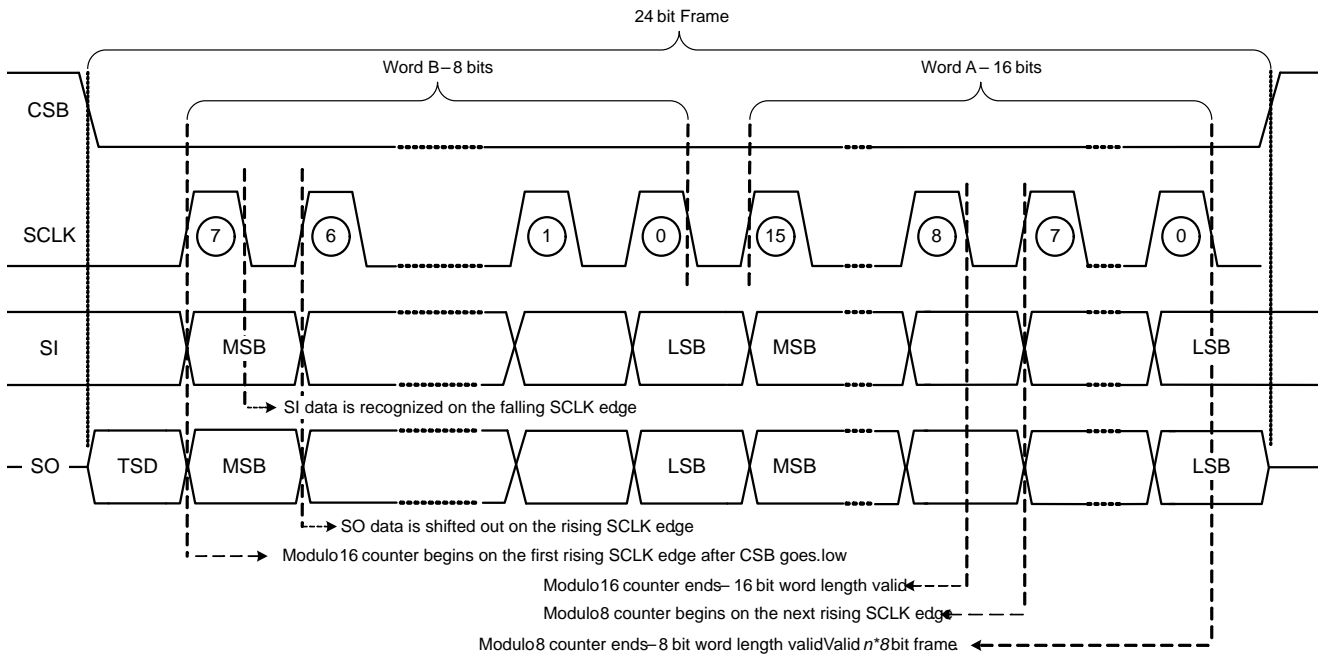


Figure 17. SPI Data Recognition and Frame Detection

The TSD bit is multiplexed with the SPI SO data and OR'd with the SI input (Figure 18) to allow for reporting in a serial daisy chain configuration in devices with the same SPI protocol. A TSD error bit as a "1" automatically propagates through the serial daisy chain circuitry from the SO output

of one device to the SI input of the next. This is shown in Figures 19 and 20; first as the daisy chained devices connected with no thermal shutdown latched fault (Figure 19) and subsequently with a TSD fault in device 1 propagating through to device 2 (Figure 20).

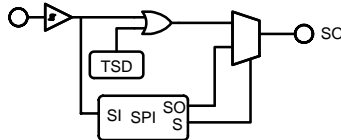


Figure 18. TSD SPI Link

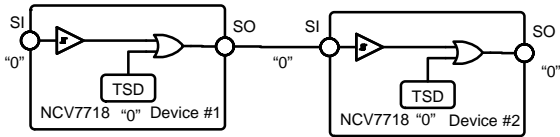


Figure 19. Daisy Chain No TSD Fault

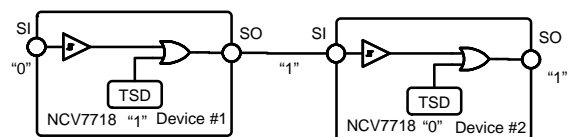


Figure 20. Daisy Chain TSD Error Propagation

DEVICE PROTECTION, DIAGNOSTICS AND FAULT REPORTING

Power Up/Down Control

Each analog power pin (VS1 or VS2) powers their respective output drivers. After a device has powered up and the output drivers are allowed to turn on, the output drivers will not turn off until the voltage on the supply pins is reduced below the initial under voltage threshold, exceeds the over voltage threshold or if shut down by either a SPI command or a fault condition.

Internal power-up circuitry on the logic supply pin supports a smooth turn on transition. VCC power up resets the internal logic such that all output drivers will be off as power is applied. All the internal counters, SI and SO along with all the digital registers will be cleared on VCC POR. Exceeding the under voltage lockout threshold on VCC allows information to be input through the SPI port for turn on control. Logic information remains intact over the entire VS1 and VS2 voltage range.

Under Voltage Shutdown

An under voltage lockout circuit prevents the output drivers from turning on unintentionally. This control is provided by monitoring the voltages on the VS1, VS2 and VCC pins. A built-in hysteresis on the under voltage threshold is included to prevent an unknown region on the power pins; VCC, VS1 and VS2. When the VCC goes below the threshold, all outputs are turned off and the input and output registers are cleared.

An under voltage condition on the VSx pins will result in shutting off all the drivers and the status bit 14 (PSF) will be set. The SPI port remains active during a VSx under-voltage if proper VCC voltage is supplied. Also all driver states will be maintained in the logic circuitry with the valid VCC voltage. Once the input voltage VSx is above the under voltage threshold level the drivers will return to programmed operation and the PSF output register bit is cleared.

Under-voltage timing diagram is provided in Figure 21.

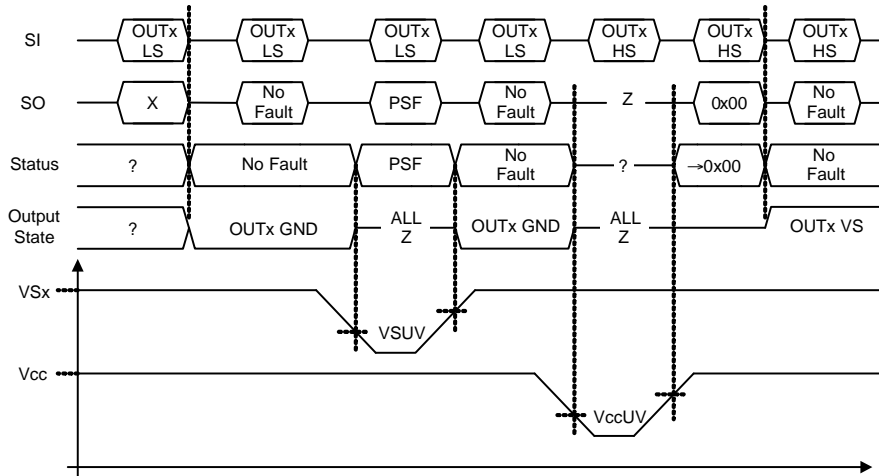


Figure 21. Under-Voltage Timing Diagram

Over Voltage Shutdown

Over voltage shutdown circuitry monitors the voltage on the VS1 and VS2 pins, which permits a 40 V maximum. When the Over-voltage Threshold level has been breached on the VS1 or VS2 supply input, the output bit 14 (PSF) will be set. Additionally, if the input bit 0 (OVLO) is asserted, all outputs will turn off. During an Over Voltage Lockout condition the turn on/off status is maintained in the logic circuitry. When proper input voltage levels are

re-established, the programmed outputs will turn back on. Over-voltage shutdown can be disabled by using the SPI input bit 0 (OVLO = 0) to run through a load dump situation. It is highly recommended to operate the part with OVLO bit asserted to ensure that the drivers remain off during a load dump scenario.

The table below describes the driver status when enabling/disabling the over voltage lock out feature during normal and overvoltage situations.

Table 3. OVER-VOLTAGE LOCK OUT (OVLO)

OVLO Input Bit	VSx OVLO Condition	Output Data Bit 14 Power Supply Fail (PSF) Status	OUTx Status
0	0	'0'	Not in Overvoltage Outputs Unchanged
0	1	'1' (Clears when VSx within Operating Range)	In Overvoltage → Outputs Unchanged
1	0	'0'	Not in Overvoltage Outputs Unchanged
1	1	'1' (Clears when VSx within Operating Range)	All Outputs Off (Remain off until VSx is out of OVLO)

Over-voltage timing diagram is provided in Figure 22.

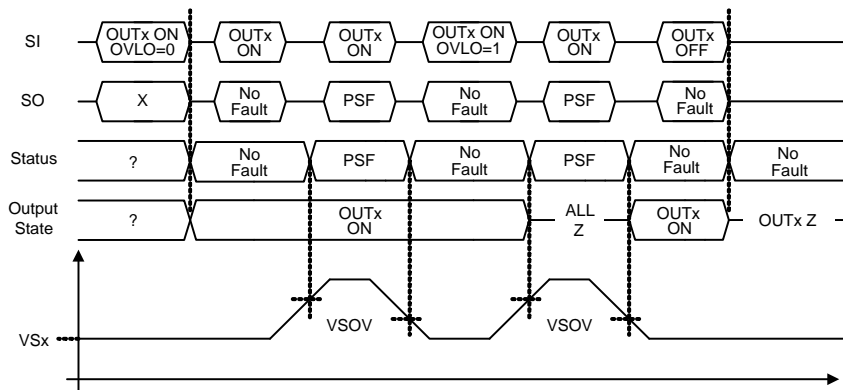


Figure 22. Over-Voltage Timing Diagram

Over Current Detection and Shutdown

The NCV7728 offers over current shutdown protection on the OUTx pins by monitoring the current on the high side and low side drivers. If the over current threshold is breached, the corresponding output is latched off (HS and LS driver is latched off) after the specified shutdown time, TdOc. Upon over current shutdown, the serial output bit OCS will be set and the corresponding HBx[1:0] will be changed to “01” to denote a high power dissipation state. Devices can be turned back on via the SPI port once the OCS condition is cleared by setting the SRR to ‘1’ on the next SPI

command. The event triggering the over current shutdown condition must be resolved prior to clearing the OCS bit to avoid repetitive stress on the drivers. Failure to do so may result in non reversible fatal damage.

The SO data OCS shown on Figure 23 corresponds to both the global SO bit #15 and the HBx OCS encoding state ‘01’.

Note: high currents could cause a high rise in die temperature. Devices will turn off if the die temperature exceeds the thermal shutdown temperature.

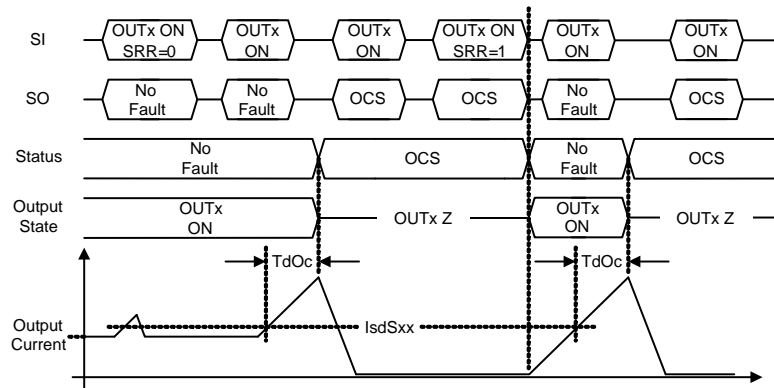


Figure 23. Over-Current Timing Diagram

Under Load Detection

The under-load detection is employed on both the HS and LS drivers. A global output bit (ULD) reports under load fault if at least one of the channels detect an under load condition. When configured for either LS or HS operation, a minimum load current (IuldLS or IuldHS) – this is the maximum open circuit detection threshold) is required when the drivers are turned on to avoid an under-load condition. If the under-load detection threshold has been breached longer than the specified under-load timer (TdUld), the ULD output bit is set to ‘1’ and HBx[1:0] bits transitions to “10”. Furthermore, if the Under-Load Detection Shutdown

Control (ULDSC bit # 13) input bit is set then the offending half-bridge output will be turned off (HS and LS drivers will be latched off). Otherwise the outputs remain active and the device configuration can be changed. The SO reporting is stored until the SRR clears the fault. Since an OCS fault has a higher priority over an under load fault, the HBx[1:0] under load fault state can be masked by an over current fault.

There is only one global under load timer for all the drivers. If the TdUld timer is already activated due to one under load, any subsequent under load delays will be the remainder of the TdUld timer.

Table 4. UNDER-LOAD DRIVER STATUS

ULDSC Input Bit 13	OUTx ULD Condition	Output Data Bit Under Load Detect Status	OUTx Status
0	0	‘0’	Unchanged
0	1	‘1’ (Need SRR to reset)	Unchanged
1	0	‘0’	Unchanged
1	1	‘1’ (Need SRR to reset)	OUTx Latches off (Need SRR to reset)

The ULD SO data provided in the under load timing diagram in Figure 24 reflects the global ULD SO bit #13 and the HBx ULD encoding state ‘10’.

NCV7728

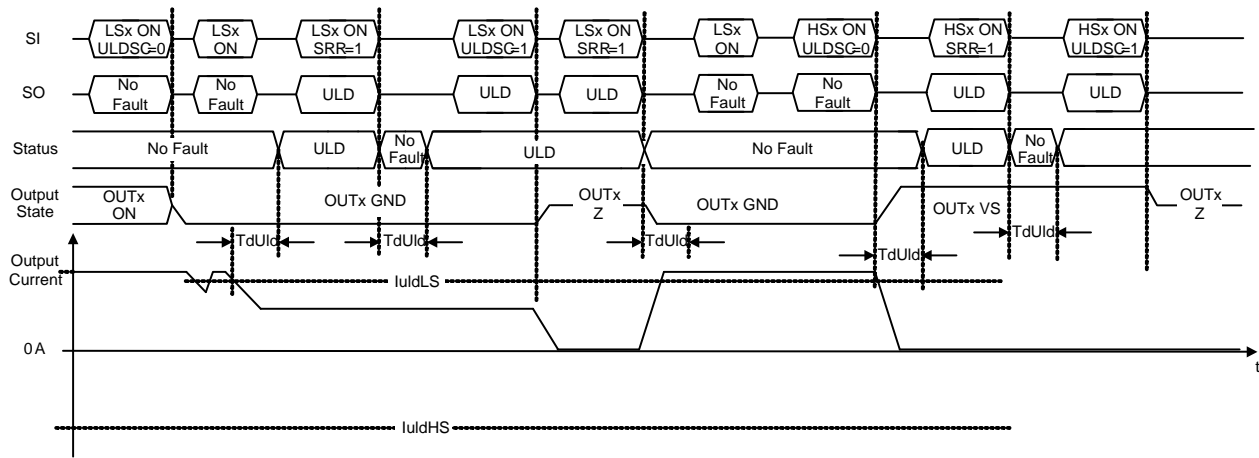


Figure 24. LS Under-load Timing Diagram

Thermal Warning and Thermal Shutdown

The NCV7728 provides individual thermal sensors for each half-bridge. Moreover, the sensor reports over temperature warning level and an over temperature shutdown level. The TW status bit (output bit 0) will be set if the temperature exceeds the over temperature warning level, but the drivers will remain active. Once the IC temperature fall below the thermal warning threshold the TW flag is automatically cleared. If any of the individual

thermal sensors detects a thermal shutdown level then the drivers on the offending half bridge are latched off. The TSD (PRE_15) bit is set to capture a thermal shutdown event. A valid SPI command with SRR and temperature below the Tsd threshold are required to clear the latched fault. Since thermal warning precedes an over temperature shutdown, software polling of this bit will allow load control and possible prevention of over temperature shutdown conditions.

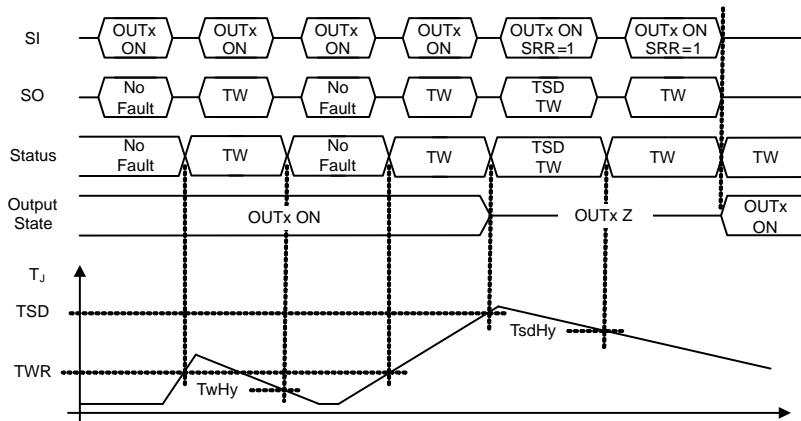


Figure 25. Thermal Warning and Shutdown Timing Diagram

Thermal Performance

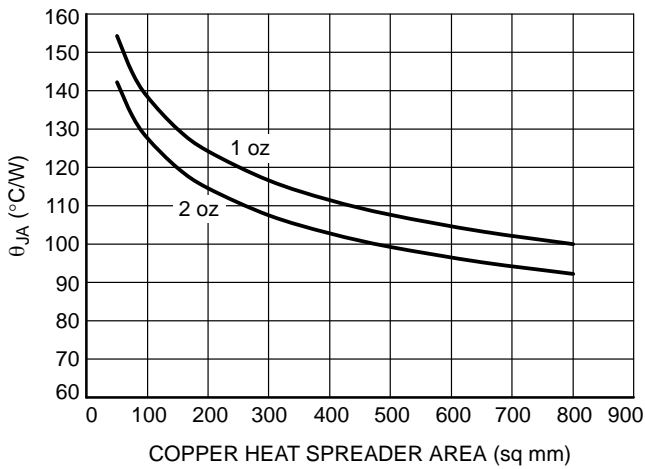


Figure 26. θ_{JA} vs. Cu Area

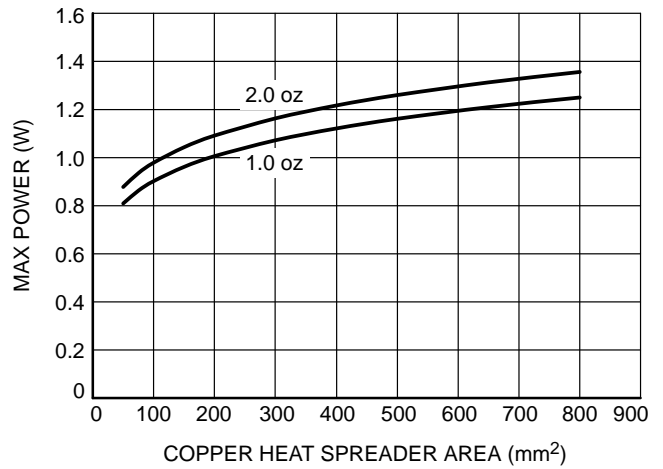


Figure 27. Power vs. Cu Area

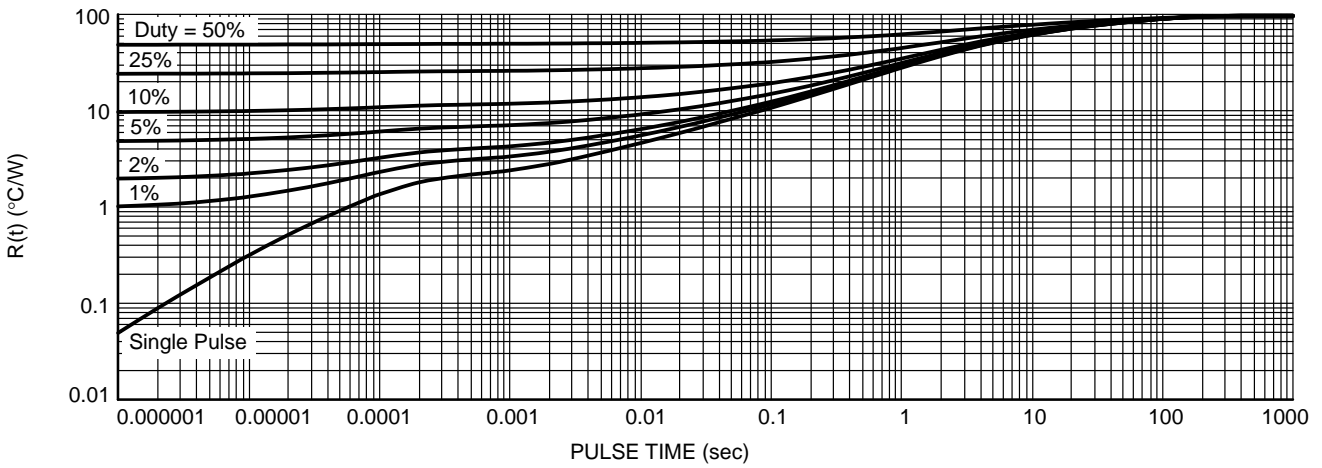


Figure 28. $R(t)$ vs. Duty Cycle

Fault Handling

At an event of a driver latched off fault, the offending half-bridge driver is disabled and the half-bridge configuration is defaulted to zero (HBENx =0, HBCNFx = 0). The user is required to clear the output register fault and to resend the proper SPI frame to turn on the drivers. A driver

that is locked out during a fault conditions auto recovers to the previous programmed state when the fault is resolved. A latched fault flag on the serial output doesn't always translate an output latched off fault.

The summary of all fault conditions, the driver status and the clear requirements are provided in Table 5.

Table 5. FAULT SUMMARY

Fault	Fault Memory Serial Output Bit	Driver Condition During Fault	Driver Condition after Parameters Within Specified Limits	Output Register Clear Requirement
Under Load (ULDSC = 0)	Latched	Outputs Unchanged. Allowed to turn/ remain on	Allowed to turn/remain on	Valid SPI frame with SRR set to 1
Under Load (ULDSC = 1)	Latched (Note 6)	Offending Half-Bridge is Latched Off (LS and HS)	Offending Half-Bridge is Latched Off (LS and HS)	Valid SPI frame with SRR set to 1
Over Current	Latched (Note 6)	Offending Output is Latched Off (LS and HS)	Offending Output is Latched Off (LS and HS)	Valid SPI frame with SRR set to 1
Thermal Warning	Non-Latched	Outputs Unchanged. Allowed to turn/ remain on provided that device is not in thermal shutdown	Allowed to turn/remain on	Temp below (thermal warning temp – hysteresis)
Thermal Shutdown	Latched (Note 6)	Offending Half-Bridge Drivers are Latched Off (LS and HS)	Offending Half-Bridge is Latched Off (LS and HS)	Valid SPI frame with SRR set to 1. Temperature blow (thermal shutdown – hysteresis)
VS Power Supply Fail (Over-Voltage: OVLO = 0)	Non-Latched	Outputs Unchanged. Allowed to turn/ remain on	Allowed to turn/remain on	VS below (Over Voltage Threshold – hysteresis)
VS Power Supply Fail (Over-Voltage: OVLO = 1)	Non-Latched	All Drivers are Locked Out. Outx → High Z	Previous Half-Bridge status and driver configuration is maintained. Allowed to turn/remain on	Auto Recovers if the VS voltage is below overvoltage threshold
VS Power Supply Fail (Under Voltage)	Non-Latched	All Drivers are Locked Out. Outx → High Z	Previous Half-Bridge status and driver configuration is maintained. Allowed to turn/remain on	Auto Recovers if the VS voltage is above the Under Voltage threshold

6. Latched conditions are cleared via the SPI SRR input bit = 1, by cycling the EN pin or with a power-on reset of V_{CC}.

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APPLICATION DIAGRAM

The application drawing below demonstrates the drive capability of the NCV7728. The VS1 and VS2 pins must be

tied together to avoid any potential difference in the supply voltage.

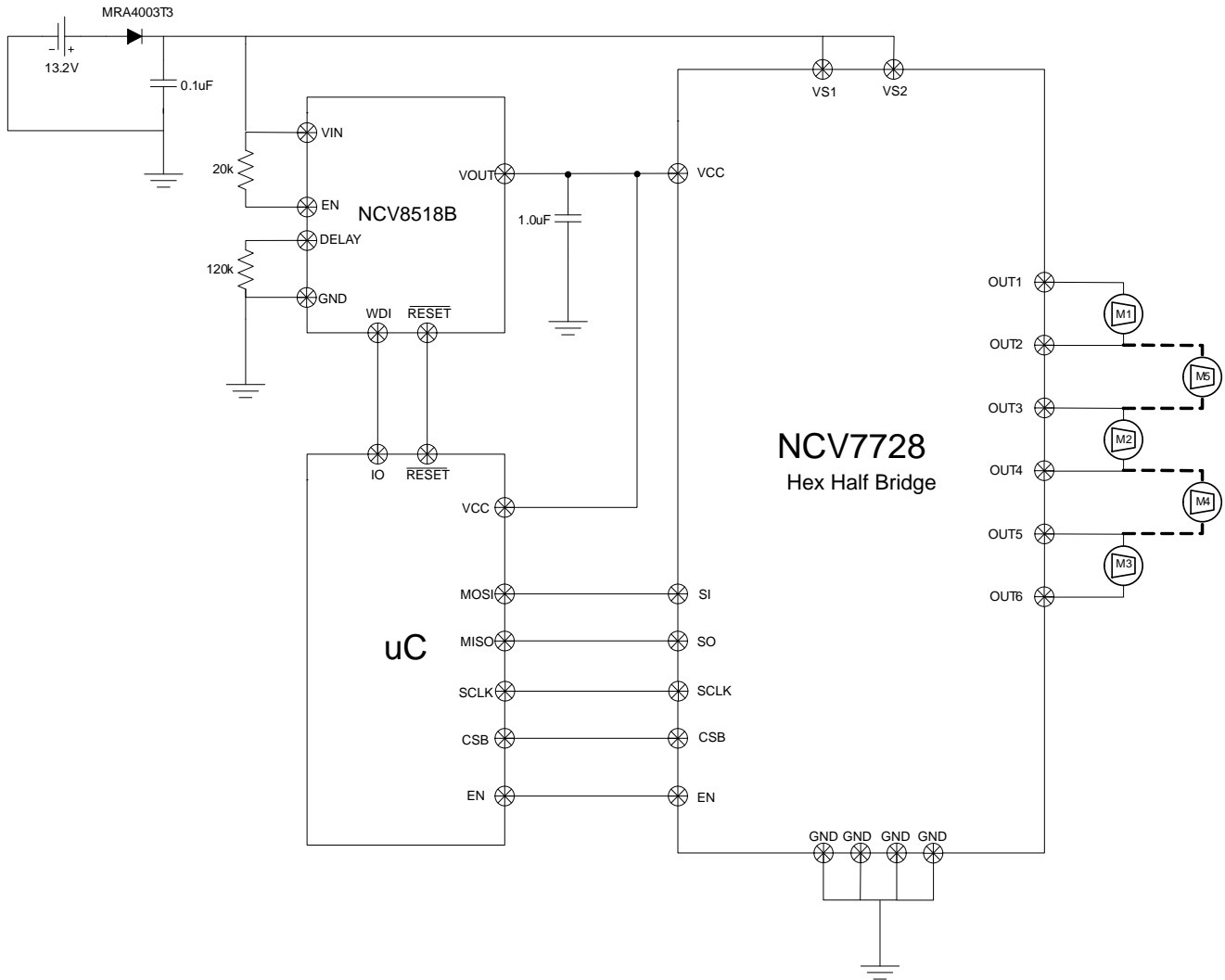


Figure 29. Application Drawing

ORDERING INFORMATION

Device	Marking	Package	Shipping†
NCV7728DPR2G	NCV7728	SSOP24 (Pb-Free)	2500 / Tape & Reel
NCV7728DPR2G	NCV7728A		

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

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